

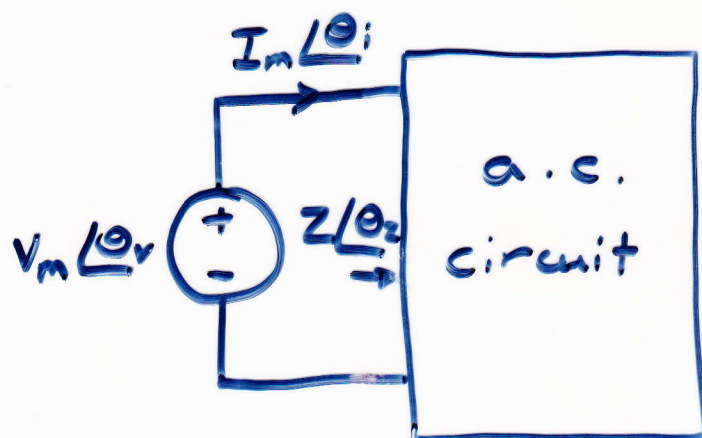
# Impedance & Admittance

## Definition

Two-terminal input IMPEDANCE,

$$\underline{Z} = \frac{\underline{V}}{\underline{I}}$$

$$\underline{Z} = \frac{V_m \angle \theta_v}{I_m \angle \theta_i} = \frac{V_m}{I_m} \angle \theta_v - \theta_i = Z \angle \theta_z$$



$$Z(\omega) = R(\omega) + jX(\omega)$$

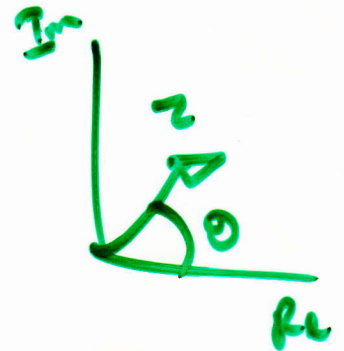
↑  
Resistive

↑  
Reactive

$$Z \angle \theta_z = R + jX$$

$$Z = \sqrt{R^2 + X^2}$$

$$\theta_z = \tan^{-1} \frac{X}{R}$$



$$R = Z \cos \theta_z$$

$$X = Z \sin \theta_z$$

Consider individual components

$$R \quad Z = R$$

$$L \quad Z = j\omega L = jX_L = \omega L \angle 90^\circ$$

$$X_L = \omega L$$

$$C \quad Z = \frac{1}{j\omega C} = \underline{jX_C} = -\frac{1}{\omega C} \angle 90^\circ$$

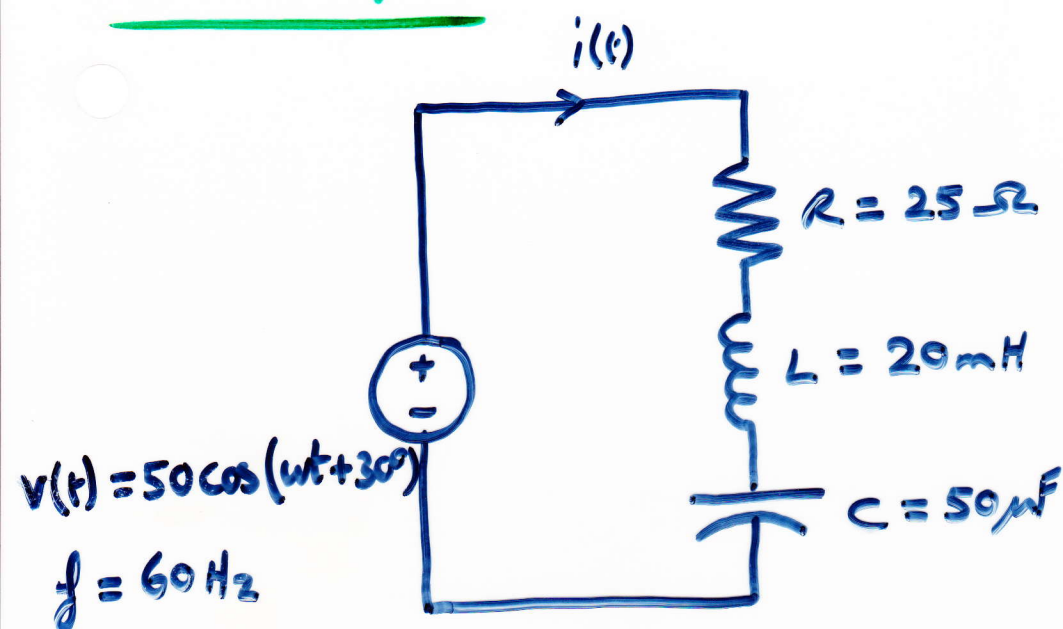
$$X_C = -\frac{1}{\omega C}$$

$$\left[ \frac{1}{j\omega C} \left( \begin{matrix} x-j \\ -j \end{matrix} \right) \right]$$

Note KCL & KVL are valid, as  
are (resistor) combination rules.

## Example

23-4



Determine

- current  $i(t)$
- equivalent impedance if  $f$  is changed to  $400 \text{ Hz}$ .

$$\underline{V} = 50 \angle 30^\circ$$

$$Z_R = 25 \Omega$$

$$Z_L = j\omega L = j(2\pi \times 60)(20 \times 10^{-3}) = j7.54 \Omega$$

$$Z_C = \frac{-j}{\omega C} = \frac{-j}{(2\pi \times 60)(50 \times 10^{-6})} = -j53.05 \Omega$$

$$\underline{Z} = \underline{Z}_R + \underline{Z}_L + \underline{Z}_C = 25 - j45.51 \Omega$$

$$= 51.93 \angle -61.22^\circ$$

$$\underline{I} = \frac{\underline{V}}{\underline{Z}} = \frac{50 \angle 30^\circ}{51.93 \angle -61.22^\circ}$$

$$= 0.96 \angle 91.22^\circ \text{ A}$$

In time domain  $i(t) = 0.96 \cos(377t + 91.22^\circ)$

If  $f = 400 \text{ Hz}$  ?

$$Z_R = 25 \Omega$$

$$Z_L = j\omega L = j50.27 \Omega$$

$$Z_C = \frac{-j}{\omega C} = -j7.96 \Omega$$

$$Z = 25 + j42.31 = 49.14 \angle 59.42^\circ \Omega$$



# Admittance

## Definition

$$\underline{Y} = \frac{1}{\underline{Z}} = \frac{\underline{I}}{\underline{V}}$$

$$\underline{Y} = Y_m \angle \theta_y$$

$$\underline{Y} = \underset{\substack{\uparrow \\ \text{conductance}}}{G} + j \underset{\substack{\uparrow \\ \text{susceptance}}}{B} = \frac{1}{R + jX}$$

Can show

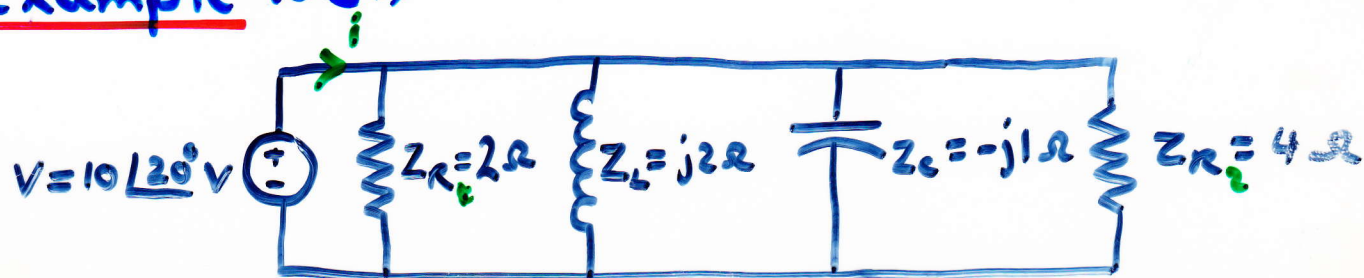
$$G = \frac{R}{R^2 + X^2}, \quad B = \frac{-X}{R^2 + X^2}$$

$$R = \frac{G}{G^2 + B^2}, \quad X = \frac{-B}{G^2 + B^2}$$

## Combining admittance

SERIES  $\frac{1}{\underline{Y}_s} = \frac{1}{\underline{Y}_1} + \frac{1}{\underline{Y}_2} + \frac{1}{\underline{Y}_3} + \dots + \frac{1}{\underline{Y}_n}$

PARALLEL  $\underline{Y}_p = \underline{Y}_1 + \underline{Y}_2 + \underline{Y}_3 + \dots + \underline{Y}_n$

Example (E89)

$$Y_{R_1} = \frac{1}{2} = 0.5 \text{ S}$$

$$Y_{R_3} = \frac{1}{4} \text{ S}$$

$$Y_L = \frac{1}{j2} = -0.5j \text{ S}$$

$$Y_C = \frac{1}{-j} = j \text{ S}$$

$$\therefore Y_p = 0.5 + 0.25 - 0.5j + j = 0.75 + 0.5j$$

$$\underline{I} = \underline{Y}_p \underline{V}$$

$$= \left( \frac{3}{4} + \frac{1}{2}j \right) \underline{V}$$

$$Y_p = \sqrt{\left(\frac{3}{4}\right)^2 + \left(\frac{1}{2}\right)^2} = \sqrt{\frac{9}{16} + \frac{1}{4}} = \sqrt{\frac{13}{16}} = 0.901$$

$$\angle Y = \tan^{-1}\left(\frac{1/2}{3/4}\right) = \tan^{-1}(2/3) = 33.7^\circ$$

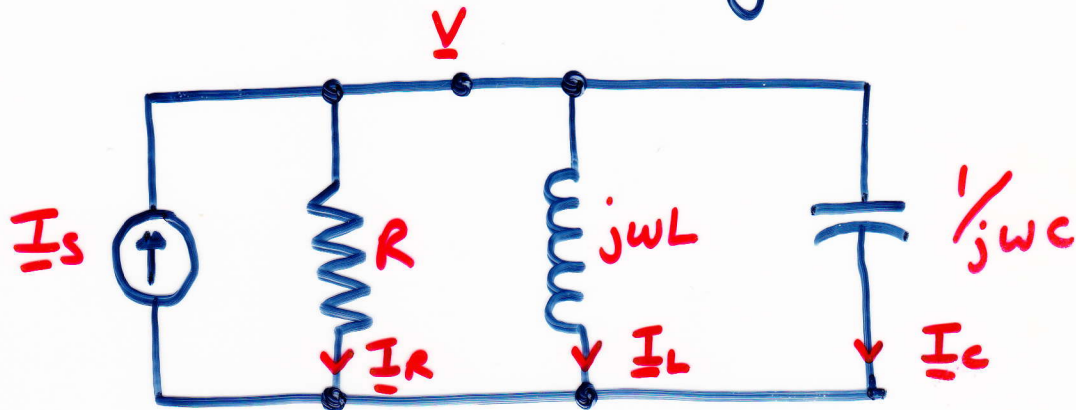
$$\therefore \underline{I} = 0.901 \angle 33.7^\circ \times 10 \angle 20^\circ$$

$$\underline{I} = 9.01 \angle 53.7^\circ \text{ A}$$

# Phasor Diagrams & Application of Kirchhoff's Laws

## Phasor Diagrams

Consider the RLC arrangement.



Let  $\underline{V}$  be a reference phasor with an arbitrarily assigned phase of  $0^\circ$ ,

$$\text{so } \underline{V} = V_m \angle 0^\circ$$

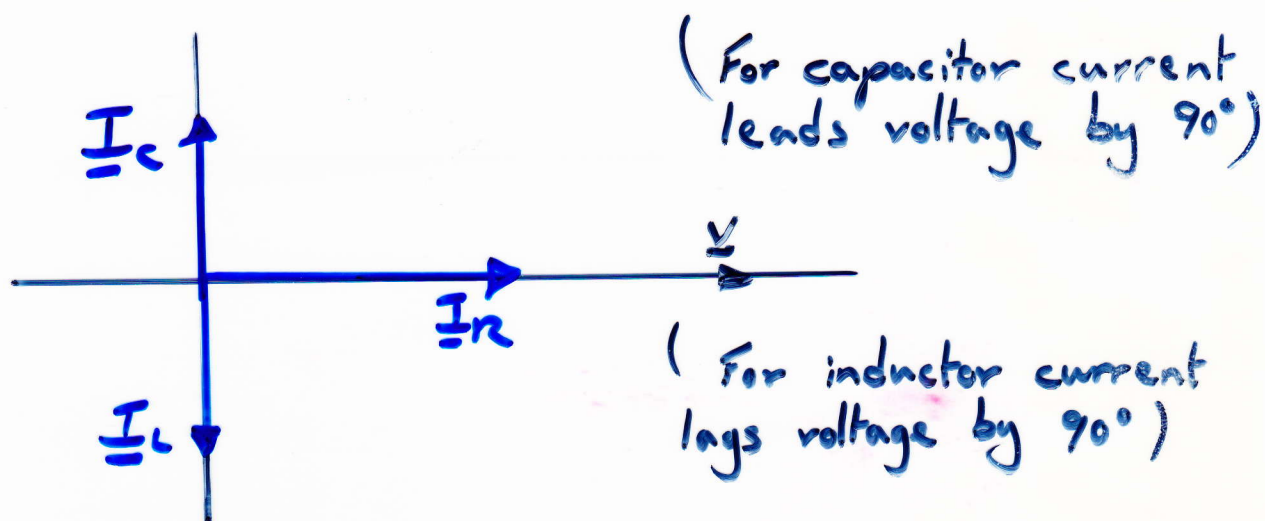
At upper node have from KCL

$$\underline{I}_s = \underline{I}_R + \underline{I}_L + \underline{I}_C = \frac{\underline{V}}{R} + \frac{\underline{V}}{j\omega L} + \frac{\underline{V}}{1/j\omega C}$$

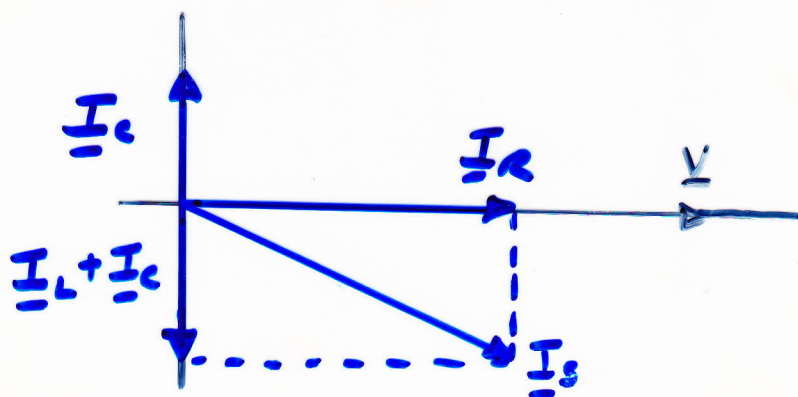


$$\therefore I_s = \frac{V_m \angle 0^\circ}{R} + \frac{V_m \angle -90^\circ}{\omega L} + V_m \omega C \angle 90^\circ$$

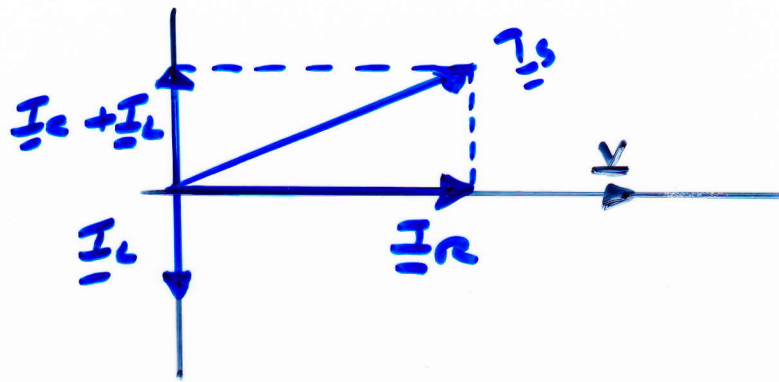
Representing these individual contributions graphically we have:



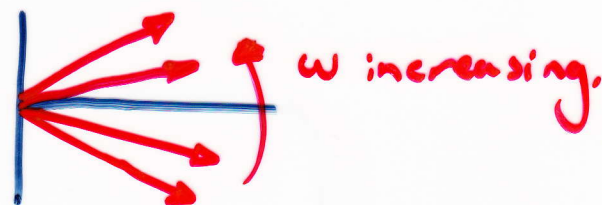
Considering relative sizes if:  
 $\omega$  is small, then  $\underline{I}_L$  is greater than  $\underline{I}_C$



$\omega$  is large, then  $\underline{I}_C$  is greater than  $\underline{I}_L$



As frequency increases phasor moves from 4th quadrant to 1st quadrant.



Now recall

$$\underline{I} = \left[ \frac{1}{R} + j(\omega C - \frac{1}{\omega L}) \right] \underline{V}$$

Can see when  $\omega C = \frac{1}{\omega L}$ , i.e.  $\underline{I}_C = \underline{I}_L$  then

$\underline{I}_S$  is in phase with  $\underline{V}$

$$\omega L = \frac{1}{\omega C}$$

$$\therefore \boxed{\omega = \frac{1}{\sqrt{LC}}}$$